## No: AnswersName: SolutionsEskişehir Osmangazi University, Faculty of Engineering and ArchitectureDepartment of Electrical Engineering & Electronics, "Communications" Final

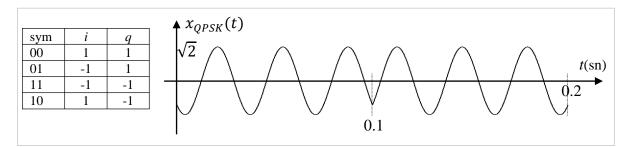
I handed over all smart devices (mobile phone, smartwatch, earphones, etc.) to the exam supervisor before the exam. I take full responsibility if any are found on me. 21.05.2025

**1.** Determine if  $x(t) = Acos(\frac{2\pi t}{T} + \varphi)$  and  $y(t) = Bcos(\frac{2\pi m t}{T} + \theta)$  are orthogonal over (0,*T*) interval, where *A*, *B*,  $\varphi$  and  $\theta$  are arbitrary constants and m = 1, 2, ... is an integer.

$$\begin{split} R &= \int_{0}^{T} A\cos\left(\frac{2\pi t}{T} + \varphi\right) B\cos\left(\frac{2\pi m t}{T} + \theta\right) dt \quad (\text{use } \cos(a)\cos(b) = \frac{1}{2}\cos(a+b) + \frac{1}{2}\cos(a-b)) \\ &= \frac{AB}{2} \int_{0}^{T} \cos\left(\frac{2\pi t}{T} + \varphi + \frac{2\pi m t}{T} + \theta\right) dt + \frac{AB}{2} \int_{0}^{T} \cos\left(\frac{2\pi t}{T} + \varphi - \frac{2\pi m t}{T} + \theta\right) dt \\ &= \frac{AB}{2} \int_{0}^{T} \cos\left(\frac{2\pi (1+m)t}{T} + \varphi + \theta\right) dt + \frac{AB}{2} \int_{0}^{T} \cos\left(\frac{2\pi (1-m)t}{T} + \varphi - \theta\right) dt \\ &= \left[\frac{ABT}{4\pi (1+m)} \sin\left(\frac{2\pi (1+m)t}{T} + \alpha\right) + \frac{ABT}{4\pi (m-1)} \sin\left(\frac{2\pi (m-1)t}{T} + \beta\right)\right]_{0}^{T} \\ &= \frac{ABT}{4\pi (1+m)} \left(\sin(2\pi (1+m) + \alpha) - \sin(\alpha)\right) + \frac{ABT}{4\pi (m-1)} \left(\sin(2\pi (m-1) + \beta) - \sin(\beta)\right) \\ &= \frac{ABT}{4\pi (1+m)} \left(\sin(\alpha) - \sin(\alpha)\right) + \frac{ABT}{4\pi (m-1)} \left(\sin(\beta) - \sin(\beta)\right) = 0? \end{split}$$

R = 0 for all integer values of m > 1; x(t) and y(t) are orthogonal. For m=1,  $|\varphi - \theta| = \frac{\pi}{2}$  must be satisfied in order to have orthogonality.

**2.** Given the IQ-table representing a QPSK constellation diagram, in-phase carrier  $c(t) = cos(60\pi t)$  and symbol rate of 10 sym/s, draw the QPSK signal for stream 1101 within 2 symbol period, marking relevant numbers on the graph (draw like an engineer!).



**3.** A memoryless source is given with symbol probabilities of  $z = \{0.1, 0.14, 0.2, 0.12, 0.12, 0.02, 0.1, 0.08, 0.07, 0.05\}$ . Estimate the possible minimum and maximum lengths of the codewords in Huffman dictionary.

 $I_{min} = -log_2(0.2) \approx 2.322$  bits, hence we can have 2 bits as minimum codeword length.  $I_{max} = -log_2(0.02) \approx 5.644$  bits, hence we can have 6 bits as maximum codeword length. Or, one may find codebook first to see min/max codeword lengths.

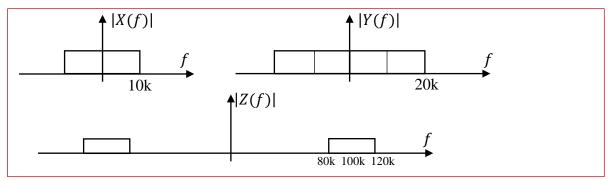
**4.** Outputs of two pn-sequence generators are modulo summed to generate another sequence. What is the length of the new sequence when SSRG[5,3] and SSRG[6,1] run with the same clock signal.

We should find smallest common period for  $N_1 = 31$  and  $N_2 = 63$ . That is, we find lowest common multiplier.  $gcd(63,31) = gcd(32,31) = gcd(1,31) = gcd(1,30) = \dots = gcd(1,1) = 1$   $lcm(63,31) = 63 \times \frac{31}{\gcd(63,31)} = 1953$ 

**5.** Calculate the entropy of the memoryless source given with symbol probabilities  $z=\{0.5, 0.25, 0.125, 0.125\}$ . Calculate the average code length when the output is compressed using Shannon-Fano dictionary method.

$$\begin{split} H(z) &= -\sum_{i=0}^{3} p_i log_2(p_i) = -0.5 \times log_2(0.5) - 0.25 \times log_2(0.25) - 0.25 \times log_2(0.125) \\ H(z) &= 1.75 \\ \text{Using Shannon-Fano method of variable length code generation, we find} \\ L &= \{1, 2, 3, 3\}. \text{ Hence } L_{avg} = \sum_{i=0}^{3} p_i l_i = 0.5 \times 1 + 0.25 \times 2 + 0.25 \times 3 = 1.75. \\ \text{This is the same as entropy, because self information of all symbols are integers.} \end{split}$$

6. A baseband signal with bandwidth of 10 kHz AM modulates a carrier with  $f_c=10$  kHz. The resulting signal is multiplied with another carrier with  $f_2=100$  kHz. What is the bandwidth of the resulting final signal?



The bandwidth of the resulting signal is 40 kHz.

7. Calculate the output noise power of a filter given as  $|H(f)| = \begin{cases} 1 - 0.5|f| & |f| < 2\\ 0 & , otherwise' \end{cases}$ when two sided input noise power spectral density is 1  $\mu$ W/Hz.

 $P_{\eta} = \int_{-\infty}^{\infty} N_0 |H(f)|^2 df = 2 \int_0^2 10^{-6} \left(\frac{f}{2}\right)^2 df = \frac{10^{-6}}{6} f^3 \Big|_0^2 = \frac{4}{3} \mu W$ 

(you get the same grade if you inadverdently calculate for one sided density)

8. Calculate the output signal power of a filter given as  $|H(f)| = \begin{cases} 1 - 0.5|f| & |f| < 2\\ 0 & |f| < 2 \end{cases}$ , otherwise, when input signal is given as  $x(t) = 2\cos(2\pi t + \pi/4) + 1$ .

Part of the signal is 1 Hz sinusoidal; the output amplitude is halved as H(1)=0.5, giving  $y(t) = cos(2\pi t + \varphi)$  from which we calculate the power as  $P_1 = 0.5 W$ . We also have a dc component whose power is  $P_2 = 1 W$ . Therefore,  $P_x = 1.5 W$ .

**9.** An integrator receiver expects the 4-ary PAM pulses of amplitudes  $\psi_i(t) = \{-4, -2, 2, 4\}$ . Find optimal threshold values at decision instants, assuming symmetric noise pdf and equal symbol probabilities.

 $R_2 = \int_0^{T_s} 2dt = 2T_s \implies R_{-2} = -2T_s, R_4 = \int_0^{T_s} 4dt = 4T_s \implies R_{-4} = -4T_s.$ The ML decision thresholds are then found as  $\{-3T_s, 0, 3T_s\}.$ 

**10.** In an OFDM system with 4  $\mu$ s symbol duration (no CP), 11 band-edge carriers are null, 1 is used for RF carrier synchronization, and the rest are used to transmit data using either BPSK, QPSK or 64-QAM. Determine the number of sub-carriers needed (FFT-size) in order to achieve 200 Mbps.

 $\begin{array}{l} 6 \times (N_c - 12)/4 \times 10^{-6} \geq 200 \times 10^6 \\ N_c \geq 200 \times 10^6 \times 4 \times 10^{-6} \ /6 + 12 \\ N_c \geq 145.3. \end{array}$